

$L \quad \underline{Z}_L = j\omega L$  

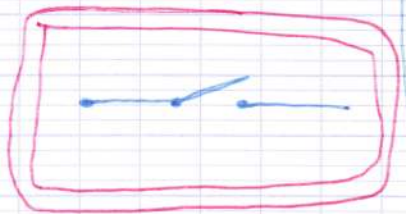
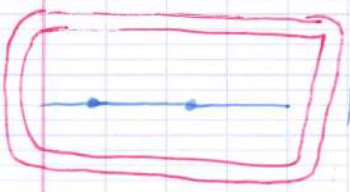
$Z_L = |\underline{Z}_L| = \omega L$


Basse Fréquence  
( $f \rightarrow 0$ )  
( $\omega \rightarrow 0$ )

Haute Fréquence  
( $f \rightarrow \infty$ )  
( $\omega \rightarrow \infty$ )

$\lim_{\omega \rightarrow 0} Z_L = 0$

$\lim_{\omega \rightarrow \infty} Z_L = +\infty$



$C \quad \underline{Z}_C = \frac{1}{j\omega C}$  

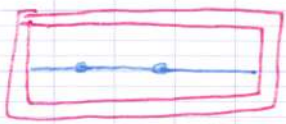
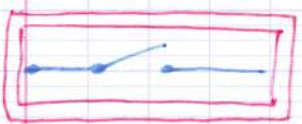
$Z_C = |\underline{Z}_C| = \frac{1}{\omega C}$

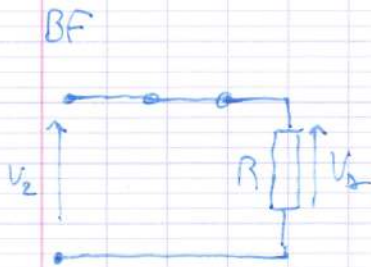
Basse Fréquence  
( $f \rightarrow 0$ )  
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Haute Fréquence  
( $f \rightarrow \infty$ )  
( $\omega \rightarrow \infty$ )

$\lim_{\omega \rightarrow 0} Z_C = +\infty$

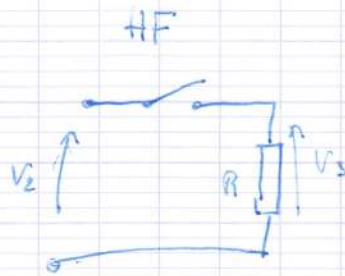
$\lim_{\omega \rightarrow \infty} Z_C = 0$





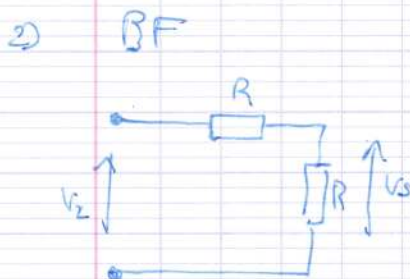
$$V_3 = V_2$$

$$A(\omega \rightarrow 0) = 1$$



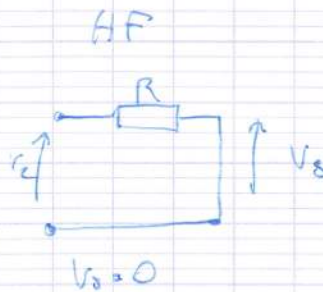
$$V_3 = 0$$

$$A(\omega \rightarrow \infty) = \frac{V_3}{V_2} = \frac{0}{V_2} = 0$$

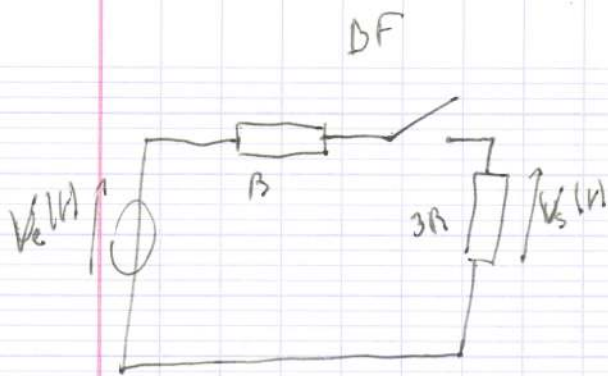


$$V_3 = \frac{V_2}{2}$$

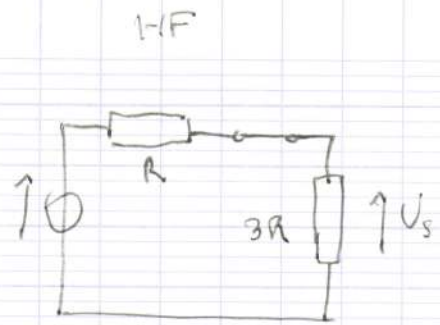
$$A(\omega \rightarrow 0) = \frac{1}{2}$$



$$A = 0$$



$V_s(t) = 0$       Filtré passe haut.



$$V_s = \frac{3R}{R+3} = \frac{3}{4} V_e(t)$$

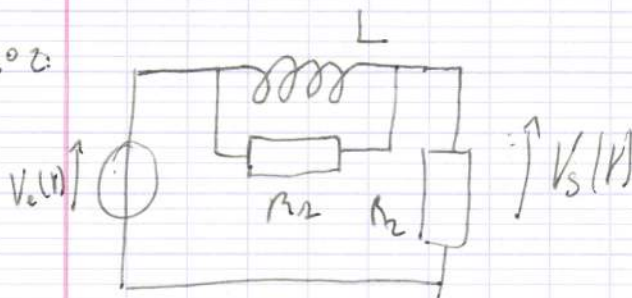
2) BF =  $\frac{5}{6}$       Filtré passe bas HF = 0

3) BF = A  $\rightarrow$  1. passe bas      HF =  $V_s(t) = \frac{2}{3} V_e(t)$

5.) BF  $V_s(t) = 0$       HF:  $V_s(t) = 0$

6) BF  $V_s(t) = 1$       HF: A  $\rightarrow$  1.      "Coupé bande."

Exercice n° 2:

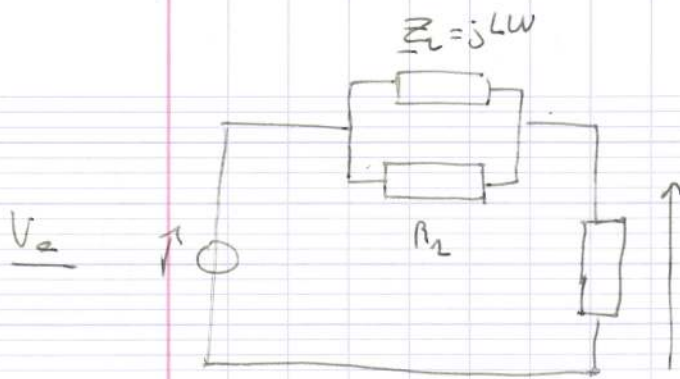


BF:  $V_s(t) = \frac{R_2}{R_2 + R_2} V_e(t)$       HF:  $V_s(t) = V_e(t)$ .  
Passe Bas.

$R_2 = 3 \text{ k}\Omega$        $G_{\text{min}} = -12 \text{ dB}$

$R_2$  A  $\rightarrow \frac{R_2}{R_2 + R_2}$  ; A  $\rightarrow$  1





$$\underline{I}(\omega) = \frac{V_s}{V_e}$$

$$\underline{V}_s \stackrel{ADJ}{=} \underline{Z} = \frac{jL R_2 \omega}{R_1 + jL\omega}$$

$$\underline{V}_s = V_e \times \frac{R_2}{R_1 + \underline{Z}}$$

$$\underline{I}(\omega) = \frac{V_s}{V_e} = \frac{R_2}{R_1 + \frac{jL R_2 \omega}{R_2 + jL\omega}} = \frac{R_2 R_2 + jL R_2 \omega}{R_2 R_2 + jL(R_1 + R_2)\omega}$$

$$= \frac{1 + j \frac{L\omega}{R_2}}{1 + jL\omega \left( \frac{R_1 + R_2}{R_1 R_2} \right)}$$

$$A(\omega) = |\underline{I}(\omega)| = \frac{\sqrt{1 + \left( \frac{L\omega}{R_2} \right)^2}}{\sqrt{1 + \left( \frac{L\omega(R_1 + R_2)}{R_1 R_2} \right)^2}}$$

$$\varphi(\omega) = \text{Arg}(\underline{I}(\omega)) = \text{Arg}(\text{numerator}) - \text{Arg}(\text{denominator}) \\ = \text{Arctan} \left( \frac{L\omega}{R_2} \right) - \text{Arctan} \left( \frac{L\omega(R_1 + R_2)}{R_1 R_2} \right)$$

$$G(\omega) = 20 \log A(\omega) = 20 \log \left( \sqrt{1 + \left( \frac{L\omega}{R_2} \right)^2} \right) - 20 \log \left( \sqrt{1 + \left( \frac{L\omega(R_1 + R_2)}{R_1 R_2} \right)^2} \right) \\ = 10 \log \left( 1 + \left( \frac{L\omega}{R_2} \right)^2 \right) - 10 \log \left( 1 + \left( \frac{L\omega(R_1 + R_2)}{R_1 R_2} \right)^2 \right)$$

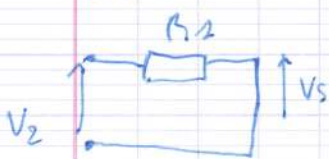
$$\lim_{\omega \rightarrow \infty} G(\omega) = \lim_{\omega \rightarrow \infty} \left[ 10 \log \left( \frac{L\omega}{R_2} \right)^2 - 10 \log \left( \frac{L\omega(R_1 + R_2)}{R_1 R_2} \right)^2 \right]$$

$$= \lim_{\omega \rightarrow \infty} \left( 20 \log \frac{\frac{L\omega}{R_2}}{\frac{L\omega(R_1 + R_2)}{R_1 R_2}} \right) = 20 \log \left( \frac{R_2}{R_1 + R_2} \right)$$

Exercice n°3 a) Passe-bas : Passe-



BF: -HF



$$\lim_{\omega \rightarrow 0} A = 0$$

$$\lim_{\omega \rightarrow \infty} A = 0$$

filtra passabanda d'ordre 2.

$$Y = \frac{1}{Z} = \frac{1}{R_2} + j\omega C + \frac{1}{j\omega L} = \frac{1}{R_2} + j\left(\omega C - \frac{1}{\omega L}\right)$$

$$\underline{T}(\omega) = \frac{V_2}{V_s} = \frac{Z}{R_2 + Z} = \frac{\frac{1}{Y}}{R_2 + \frac{1}{Y}} = \frac{1}{1 + R_2 Y}$$

$$\underline{T}(\omega) = \frac{1}{1 + R_2 \left( \frac{1}{R_2} + j\left(\omega C - \frac{1}{\omega L}\right) \right)}$$

$$A(\omega) = |\underline{T}(\omega)| = \frac{1}{\sqrt{\left(1 + \frac{R_2}{R_2}\right)^2 + \left(jR_2\left(\omega C - \frac{1}{\omega L}\right)\right)^2}}$$

$$\varphi(\omega) = \text{Arg}(\underline{T}(\omega)) = 0 - \text{Arctan}\left[\frac{R_2\left(\omega C - \frac{1}{\omega L}\right)}{1 + \frac{R_2}{R_2}}\right]$$

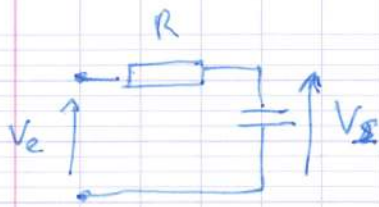
$$G(\omega) = 20 \log A(\omega)$$

$$G_{\max} = 20 \log A_{\max}$$

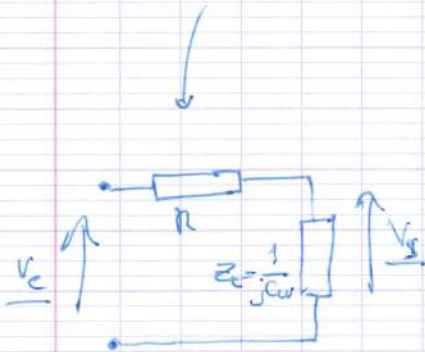
$$A_{\max} \text{ quand } \omega C - \frac{1}{\omega L} = 0 \Leftrightarrow \omega C = \frac{1}{\omega L}$$

$$\boxed{LC\omega^2 = 1}$$





$$\underline{T(\omega)} = \frac{V_s}{V_e} = \frac{1}{1 + RC\omega}$$



$$A(\omega) = |T(\omega)| = \frac{V_s}{V_e} = \frac{1}{\sqrt{1 + (RC\omega)^2}}$$

$$\phi(\omega) = \theta_{V_s} - \theta_{V_e} = -\text{Arctan}(RC\omega)$$

$$G(\omega) = 20 \log(A(\omega)) =$$

$$20 \log \frac{1}{\sqrt{1 + (RC\omega)^2}} = -10 \log(1 + (RC\omega)^2)$$

$$V_s = \frac{\frac{1}{j\omega C}}{\frac{1}{j\omega C} + R} \cdot V_e$$

Coupe/Passe bande: Fréquences limites  $G_{MAX}$  ou  $G_{MIN}$

Coupe Haut/Bas: Fréquence coupure.

$$A(\omega_c) = \frac{A_{max}}{\sqrt{2}} \rightarrow \frac{1}{\sqrt{1 + (RC\omega_c)^2}} = \frac{1}{\sqrt{2}} \Rightarrow \omega_c = \frac{1}{RC}$$

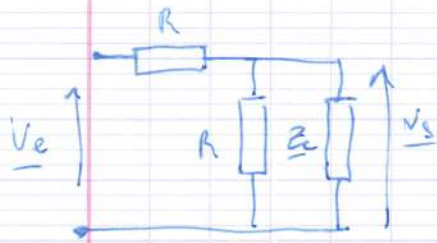
$$\underline{T(\omega)} = \frac{1}{1 + jRC\omega} = A_{max} \frac{1}{1 + j\frac{\omega}{\omega_c}} \quad \left. \begin{array}{l} \text{Forme normalisée} \\ \text{filtre passe bas du 1}^{\text{er}} \text{ ordre} \end{array} \right\}$$

$A_{max} = 1 \quad G_{max} = 0 \text{ dB}$

Forme normalisée

$$A(\omega) = A_{MAX} \cdot \frac{1}{1 + j\frac{\omega}{\omega_c}}$$

Passe bas 1<sup>er</sup> ordre:  $\underline{T}(\omega) = \frac{1}{1 + j\frac{\omega}{\omega_c}} \cdot A_{MAX}$



$$Y = \frac{1}{R} + j\omega C$$

$$I(\omega) = \frac{V_s}{2 + R \cdot Y} = \frac{1}{1 + j\omega RC}$$

$$T(\omega) = \frac{1 \cdot \frac{V_s}{2}}{2 + j\omega RC}$$

$$T(\omega) = \frac{\frac{1}{2}}{1 + j\frac{\omega RC}{2}} = \frac{A_{MAX}}{1 + j\frac{\omega}{\omega_c}}$$

$$A(\omega) = \frac{1}{\sqrt{4 + (\omega RC)^2}} \quad A_{MAX} = \frac{1}{2} \quad A(\omega_c) = \frac{1}{\sqrt{2}}$$

$$G(\omega) = -10 \log(4 + (\omega RC)^2)$$

$$\varphi(\omega) = -\text{Arctan}\left(\frac{\omega RC}{2}\right)$$

$$\frac{1}{\sqrt{4 + (\omega RC)^2}} = \frac{1}{2\sqrt{2}} = \frac{1}{\sqrt{8}}$$

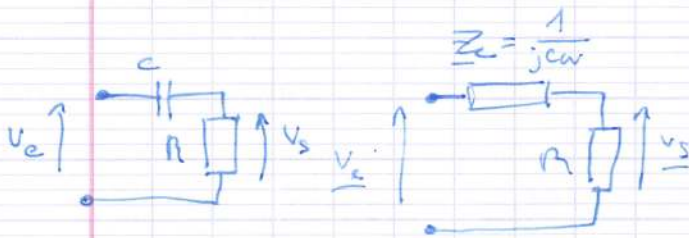
$$\omega_c = \frac{2}{RC}$$

$$T(\omega) = \frac{5R}{6R + j\omega L} = \frac{\frac{5}{6}}{2 + j\frac{\omega L}{6R}}$$

$$A_{MAX} = \frac{5}{6}$$

$$\omega_c = \frac{6R}{L}$$

Boxe Haut du 1<sup>er</sup> ordre :  $T(\omega) = A_{MAX} \cdot \frac{j\omega}{1 + j\frac{\omega}{\omega_c}}$



$$V_s = \frac{R}{R + \frac{1}{j\omega C}} \cdot V_e$$

$$T(\omega) = \frac{V_s}{V_e} = \frac{R}{R + j\frac{1}{\omega C}} = \frac{1}{1 + j\left(\frac{1}{\omega CR}\right)} = \frac{A_{MAX} = 1}{1 + j\frac{\omega}{\omega_c}}$$

$$\frac{j\omega CR}{1 + j\omega CR} = \frac{j\frac{\omega}{\omega_c}}{1 + j\frac{\omega}{\omega_c}}$$

$$\omega_c = \frac{1}{4RC} = \frac{3}{4} \times \frac{j\omega}{1 + j\frac{\omega}{\omega_c}}$$



Correction QCM:

$$\frac{1}{C\omega} = \frac{V}{I} = \frac{\frac{20}{\sqrt{2}}}{\frac{5 \cdot 10^{-3}}{\sqrt{2}}} = 4 \times 10^3 = 0,25 \mu\text{F}.$$

4) VRAI 5) FAUX.

fonction du 1<sup>er</sup> ordre:  $T(\omega) = A_{\text{max}} \cdot \frac{1}{1 + j \frac{\omega}{\omega_c}}$

$$A(\omega) = A_{\text{max}} \cdot \frac{1}{\sqrt{1 + \left(\frac{\omega}{\omega_c}\right)^2}}$$

$$G(\omega) = 20 \log \frac{A_{\text{max}}}{\sqrt{1 + \left(\frac{\omega}{\omega_c}\right)^2}} = 20 \log(A_{\text{max}}) - 10 \log\left(1 + \left(\frac{\omega}{\omega_c}\right)^2\right)$$

$$\varphi(\omega) = -\text{Arctan}\left(\frac{\omega}{\omega_c}\right)$$

En BF:  $G(\omega) = 20 \log \frac{A_{\text{max}}}{\sqrt{1 + \left(\frac{\omega}{\omega_c}\right)^2}}$

( $\omega < \omega_c$ )

devenir petit devant 1

⇒ Asymptote horizontale:  $20 \log(A_{\text{max}}) = G_{\text{max}}$ .

En HF:  $G(\omega) = 20 \log \frac{A_{\text{max}}}{\sqrt{1 + \left(\frac{\omega}{\omega_c}\right)^2}}$

( $\omega > \omega_c$ )

$$1 + \left(\frac{\omega}{\omega_c}\right)^2 \approx \left(\frac{\omega}{\omega_c}\right)^2$$

⇒ Asymptote:  $20 \log\left(A_{\text{max}} \cdot \frac{\omega_c}{\omega}\right)$ .

$$20 \log(A_{\text{max}}) + 20 \log\left(\frac{\omega_c}{\omega}\right) = 20 \log(A_{\text{max}}) - 20 \log\left(\frac{\omega}{\omega_c}\right)$$

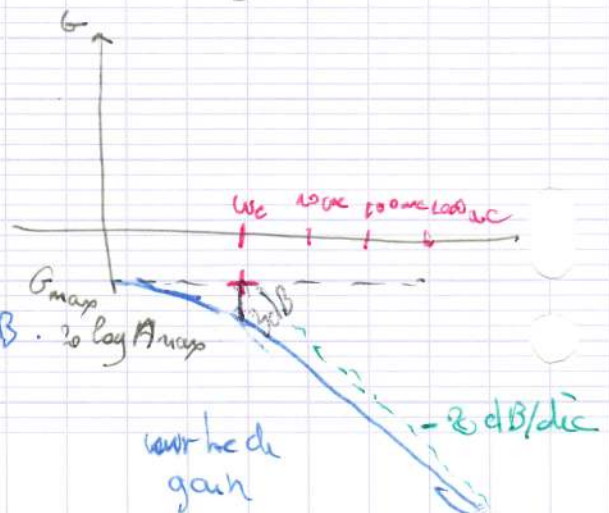
$\omega = 10 \omega_c$ :  $G \approx G_{\text{max}} - 20 \text{ dB}$

$\omega = 100 \omega_c$ :  $G \approx G_{\text{max}} - 40 \text{ dB}$

$\omega = 1000 \omega_c$ :  $G \approx G_{\text{max}} - 60 \text{ dB}$

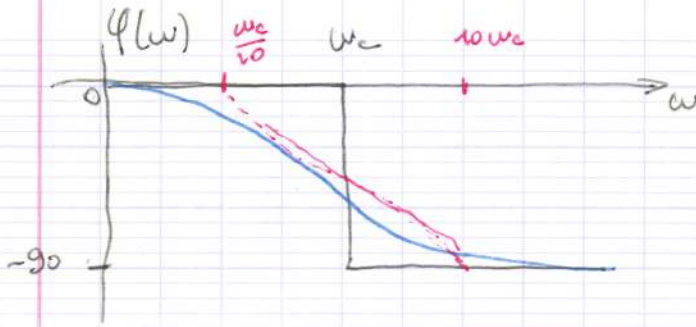
$\omega = 10000 \omega_c$ :  $G \approx G_{\text{max}} - 80 \text{ dB}$

$\omega = 10^n \omega_c$ :  $G \approx G_{\text{max}} - n \times 20 \text{ dB}$





11/04  
Electro



$$\lim_{\omega \rightarrow \infty} 20 \log \frac{A_{max}}{\sqrt{1 + \left(\frac{\omega}{\omega_c}\right)^2}} = 20 \log \frac{A_{max}}{\left(\frac{\omega}{\omega_c}\right)} = 20 \log \frac{A_{max}}{\omega/\omega_c}$$

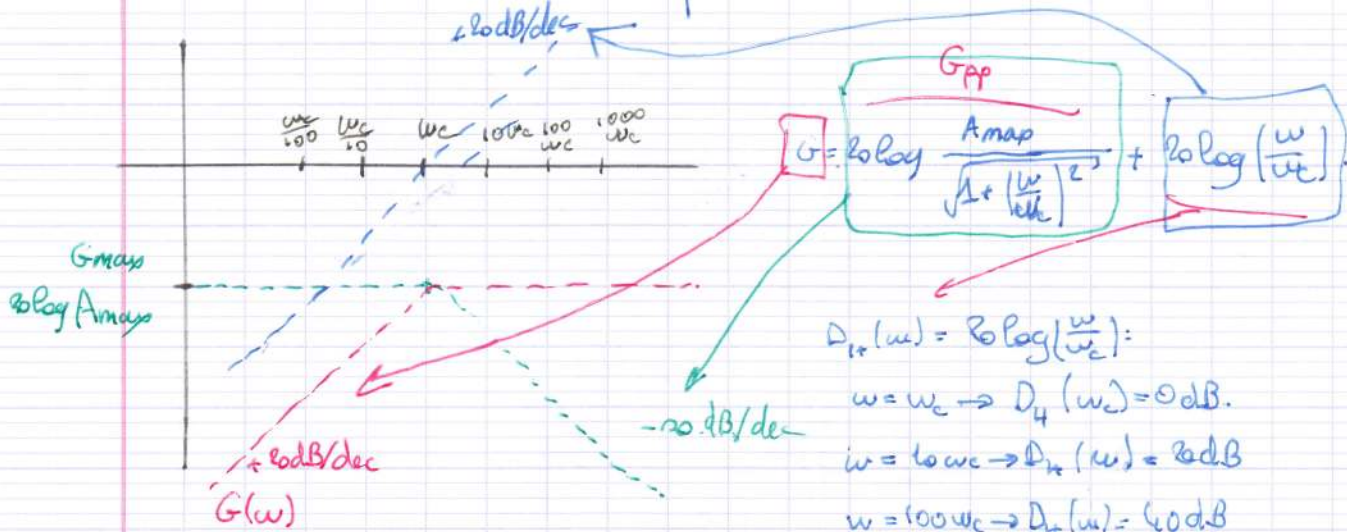
Passe Haut du 1<sup>er</sup> ordre:  $T(\omega) = A_{max} \cdot \frac{j\omega/\omega_c}{1 + j\omega/\omega_c}$

$$A(\omega) = A_{max} \frac{\omega/\omega_c}{\sqrt{1 + \left(\frac{\omega}{\omega_c}\right)^2}}$$

$$\phi(\omega) = \frac{\pi}{2} - \text{Arctan}\left(\frac{\omega}{\omega_c}\right) \quad (\text{rad})$$

$$G(\omega) = 20 \log A_{max} \frac{\omega/\omega_c}{\sqrt{1 + \left(\frac{\omega}{\omega_c}\right)^2}}$$

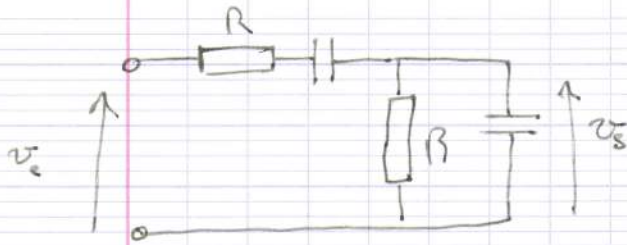
$$\phi(\omega) = 90 - \text{Arctan}\left(\frac{\omega}{\omega_c}\right) \quad (^\circ)$$



$$D_H(\omega) = 20 \log\left(\frac{\omega}{\omega_c}\right):$$

- $\omega = \omega_c \rightarrow D_H(\omega) = 0 \text{ dB}$
- $\omega = 10\omega_c \rightarrow D_H(\omega) = 20 \text{ dB}$
- $\omega = 100\omega_c \rightarrow D_H(\omega) = 40 \text{ dB}$
- $\omega = \frac{\omega_c}{10} \rightarrow D_H(\omega) = -20 \text{ dB}$

Couleur verte: courbe bleue = couleur rouge.



$$\underline{T}(\omega) = \frac{1}{1 + (R + \frac{1}{j\omega C}) \underline{Y}}$$

$$\underline{Y} = \frac{1}{R} + j\omega C$$

$$\underline{T}(\omega) = \frac{1}{1 + (R + \frac{1}{j\omega C})(\frac{1}{R} + j\omega C)} = \frac{1}{1 + 1 + j\omega RC + \frac{1}{j\omega RC} + 1}$$

$$\underline{T}(\omega) = \frac{1}{3 + j\omega RC + \frac{1}{j\omega RC}} = \frac{j\omega RC}{1 + 3j\omega RC - (\omega RC)^2}$$

Filtre du 2<sup>e</sup> ordre:

$$\underline{T}(\omega) = \frac{N(\omega)}{D(\omega)} = \frac{1}{1 + j3 \cdot \frac{\omega}{\omega_c} - (\frac{\omega}{\omega_c})^2}$$

Passer bas

$$\underline{T}(\omega) = \frac{A_0}{1 + 2j3 \frac{\omega}{\omega_c} + (\frac{\omega}{\omega_c})^2}$$

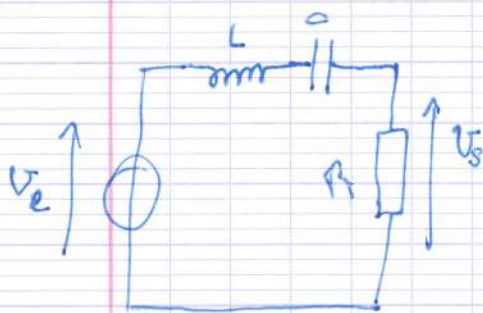
Passer bande

$$\underline{T}(\omega) = \frac{A_0 \cdot j \frac{\omega}{\omega_c}}{1 + 2j3 \frac{\omega}{\omega_c} - (\frac{\omega}{\omega_c})^2}$$

Passer haut

$$\underline{T}(\omega) = \frac{A_0 (-\frac{\omega}{\omega_c})^2}{1 + 2j3 \frac{\omega}{\omega_c} - (\frac{\omega}{\omega_c})^2}$$

$$\underline{T}(\omega) = \frac{j\omega L}{R + j\omega L} = \frac{j \frac{\omega L}{R}}{1 + j \frac{\omega L}{R}}$$



BFE:

$$\lim_{\omega \rightarrow 0} = 0$$

$$\lim_{\omega \rightarrow \infty} = v_e$$

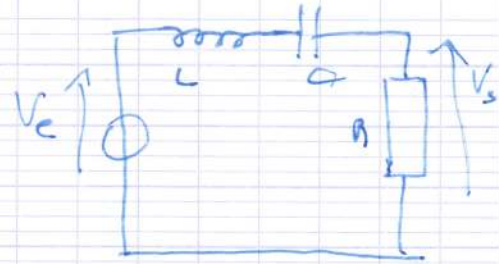


$$\textcircled{1} \quad \underline{T}(\omega) = A_0 \frac{\text{Num}(\omega)}{1 + 2j \frac{\omega}{\omega_0} - \left(\frac{\omega}{\omega_0}\right)^2}$$

$\text{Num}(\omega) = 1 \Rightarrow$  Filtre passe bas

$\text{Num}(\omega) = 2j \frac{\omega}{\omega_0} \Rightarrow$  Filtre passe bande

$\text{Num}(\omega) = -\left(\frac{\omega}{\omega_0}\right)^2 \Rightarrow$  Filtre passe haut.



$$\underline{T}(\omega) = \frac{jR\omega}{1 + jR\omega - L\omega^2}$$

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

Filtre passe bande:

$$\text{Num}(\omega) = jR\omega = 2j \frac{\omega}{\omega_0}$$

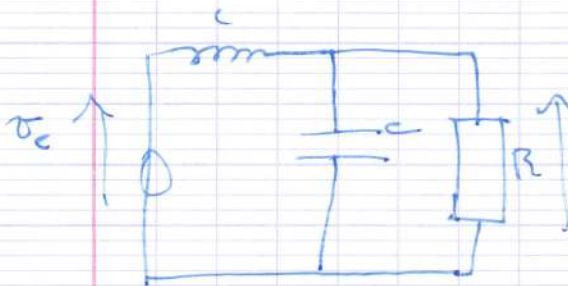
$$\text{et } \left(\frac{\omega}{\omega_0}\right)^2 = LC\omega^2$$

$$\text{et } 2j \frac{\omega}{\omega_0} = jR\omega$$

$$A_0 = 1$$

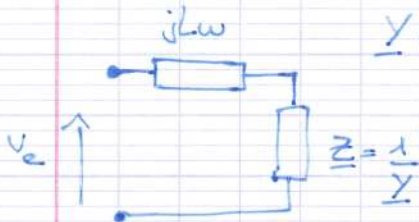
$$Z = \frac{R}{2} \sqrt{\frac{C}{L}} \leftarrow \frac{RC}{2\sqrt{LC}}$$

②



$$\underline{T}(\omega) = A_0 \frac{\text{Num}(\omega)}{1 + 2j \frac{\omega}{\omega_0} - \left(\frac{\omega}{\omega_0}\right)^2}$$

$$\underline{T}(\omega) =$$



$$Y = \frac{1}{R} + jC\omega$$

$$Z = \frac{R \times \frac{1}{jC\omega}}{R + \frac{1}{jC\omega}} = \frac{R}{1 + jR\omega C}$$

$$\underline{V}_s = \frac{Z}{Z + jL\omega} \underline{V}_e$$

$$\underline{T}(\omega) = \frac{V_s}{V_e} = \frac{Z}{Z + jL\omega} = \frac{1}{1 + \frac{jL\omega}{Z}} = \frac{1}{2 + jL\omega \times Y}$$

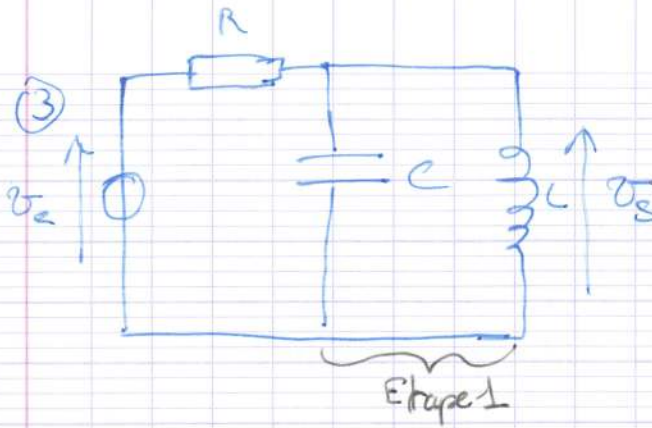
$$\underline{T}(\omega) = \frac{1}{1 + jL\omega \left(\frac{1}{R} + jC\omega\right)} = \frac{1}{1 + j \frac{L\omega}{R} - LC\omega^2} = \frac{A_0}{1 + 2j \frac{\omega}{\omega_0} - \left(\frac{\omega}{\omega_0}\right)^2}$$

$$A_0 = 1$$

$$\omega_0 = \frac{1}{\sqrt{LC}}$$

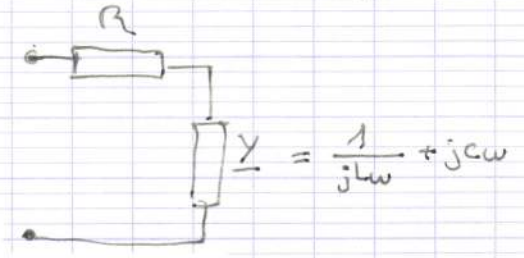
$$Z = \frac{1}{2R} \times \sqrt{\frac{L}{C}}$$

$$\begin{cases} A_0 = 1 \\ \frac{L\omega}{R} = 2j \left(\frac{\omega}{\omega_0}\right) \\ LC\omega^2 = \left(\frac{\omega}{\omega_0}\right)^2 \end{cases}$$



⇒ Passe bande d'ordre 2 :

$$\underline{T}(\omega) = \frac{2j\zeta \frac{\omega}{\omega_0}}{1 + 2j\zeta \frac{\omega}{\omega_0} - \left(\frac{\omega}{\omega_0}\right)^2}$$



$$\underline{T}(\omega) = \frac{1}{1 + R\left(\frac{1}{jL\omega} + jC\omega\right)} =$$

$$\underline{T}(\omega) = \frac{1}{1 + \frac{R}{jL\omega} + jRC\omega} = \frac{jL\omega}{jL\omega + R + j^2 LRC\omega^2}$$

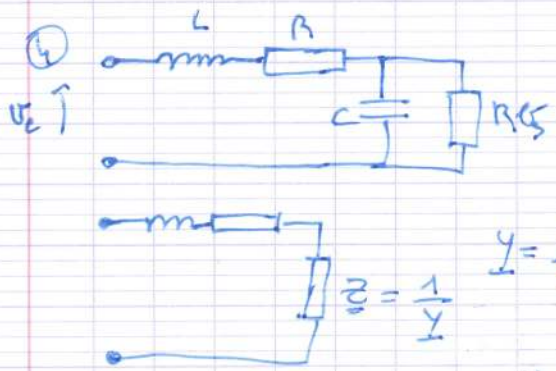
$$\underline{T}(\omega) = \frac{jL\omega}{R + jL\omega - LRC\omega^2}$$

$$\underline{T}(\omega) = \frac{jL\omega}{R} \frac{1}{1 + \frac{jL\omega}{R} - L\omega^2}$$

$$\underline{T}(\omega) = A_0 \frac{2j\zeta \frac{\omega}{\omega_0}}{1 + 2j\zeta \frac{\omega}{\omega_0} - \left(\frac{\omega}{\omega_0}\right)^2}$$

$$\begin{cases} LC = \frac{1}{\omega_0^2} \\ \frac{2\zeta}{\omega_0} = \frac{L}{R} \\ A_0 \cdot 2j\zeta \frac{\omega}{\omega_0} = j \frac{L\omega}{R} \end{cases}$$

$$\begin{cases} \omega_0 = \frac{1}{\sqrt{LC}} \\ \zeta = \frac{L\omega_0}{2R} = \frac{1}{2R\sqrt{C}} \\ A_0 = \frac{L}{R} \times \frac{\omega_0}{L} = \frac{1}{R\sqrt{LC}} = 2R\sqrt{\frac{C}{L}} \\ A_0 = 1 \end{cases}$$



$\lim_{\omega \rightarrow \infty} = 2R$  passe Bas -  
 $\lim_{\omega \rightarrow 0} = 0$

$$Y = \frac{1}{R} + jC\omega \quad Z = \frac{R \times \frac{1}{jC\omega}}{R + \frac{1}{jC\omega}} = \frac{R}{1 + RjC\omega}$$





$$\underline{I}(w) = \frac{1}{1 + (jLw + R) \times (jCw + \frac{1}{R})} = \frac{1}{1 - LCw^2 + \frac{jLw}{R} + BjCw + 1}$$

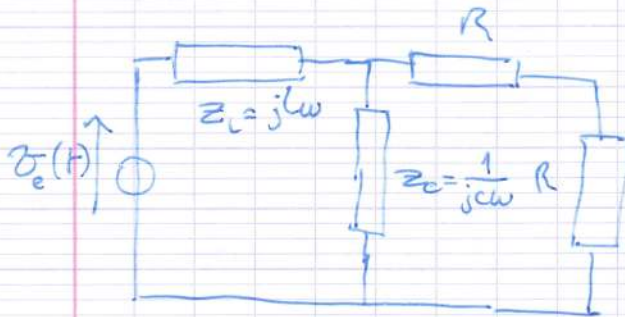
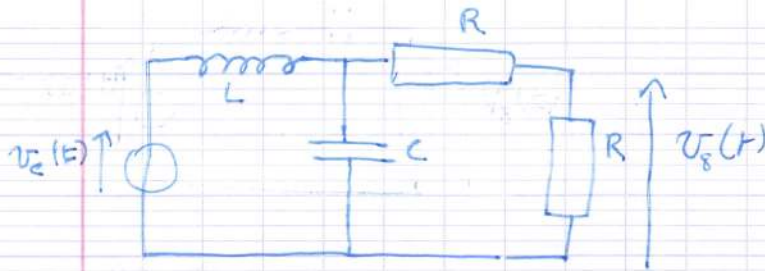
$$= \frac{1/2}{1 + \frac{jLw + R^2jCw}{2R} - \frac{LCw^2}{2}}$$

$j \times j = -1$

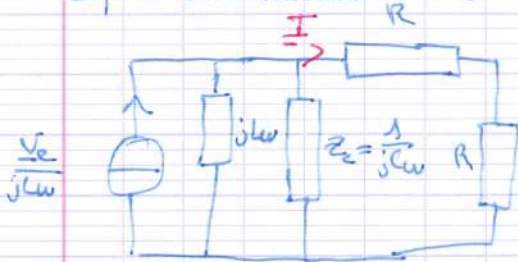
$$\left\{ \begin{array}{l} A_0 = \frac{1}{2} \\ \frac{LCw^2}{2} = \frac{w^2}{w_0^2} \\ \frac{jLw + R^2jCw}{2R} = \end{array} \right.$$

Prof à la

Electro  
16/05



Equivalence Thévenin - Norton :



$$v_s = R \cdot I$$

$$I = \frac{\frac{1}{2R}}{\frac{1}{2R} + j\omega C + \frac{1}{j\omega C}} \times \frac{v_e}{j\omega C}$$

$$v_s = R \times \frac{1}{2R} \times \frac{1}{\frac{1}{2R} + j\omega C + \frac{1}{j\omega C}} \times \frac{v_e}{j\omega C}$$

$$v_s = \frac{\frac{1}{2}}{\frac{1}{2R} + j\omega C + \frac{1}{j\omega C}} \times \frac{v_e}{j\omega C}$$

$$\frac{v_s}{v_e} = \frac{\frac{1}{2}}{\frac{j\omega C}{2R} + L\omega^2 + 1}$$

$$I(\omega) = \frac{v_s}{v_e} = \frac{\frac{1}{2}}{\frac{j\omega C}{2R} + L\omega^2 + 1}$$

$$= \frac{1}{2} \times \frac{1}{1 + \frac{j\omega C}{2R} - L\omega^2}$$

$$\begin{cases} L\omega^2 = \left(\frac{\omega}{\omega_0}\right)^2 \\ \frac{L\omega}{2R} = 2\sigma \frac{\omega}{\omega_0} \\ A_0 = \frac{1}{2} \end{cases} \Leftrightarrow \begin{cases} CL = \frac{1}{\omega_0^2} \\ L\omega_0 = 4R\sigma \\ A_0 = \frac{1}{2} \end{cases}$$

$$\Leftrightarrow \begin{cases} \omega_0 = \frac{1}{\sqrt{CL}} \\ \sigma = \frac{L}{4R\sqrt{CL}} = \frac{1}{4R} \sqrt{\frac{L}{C}} \\ A_0 = \frac{1}{2} \end{cases}$$

Filtre passe-bas :

$$= A_0 \times \frac{2}{1 + 2j\sigma \frac{\omega}{\omega_0} - \left(\frac{\omega}{\omega_0}\right)^2}$$

→ Circuit linéaire au régime sinusoïdale

→ 1<sup>er</sup> et 2<sup>e</sup> ordre

Diagramme de Bode du premier ordre